

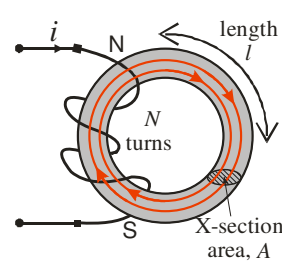
Magnetic circuits

- definition (introduction)
- electric circuit analogy
- composite magnetic circuits
- inductance
 - on dc
 - on ac
- inductor (real)
 - power loss
 - equivalent circuit
 - phasor diagram
- limitations of magnetic circuit model

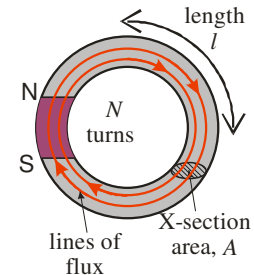
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Introduction

- ⇒ a characteristic of lines of magnetic flux is that each line a closed path
- ⇒ the complete closed path followed by any group of lines of magnetic flux is referred to as a **magnetic circuit**



•Coil current as source of mmf, F



•PM as source of mmf, F

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Electric circuit analogy

- ⇒ in electric circuits: electric current is due to effect of emf
- ⇒ in magnetic circuits: magnetic flux is due to effect of mmf
- ⇒ if a current-carrying coil is used as source of mmf, the mmf F generated when a current i flow in coil on N turns is

$$F = Ni$$

- ⇒ the mmf per unit length of the magnetic circuit is called **magnetic field strength, H**
- ⇒ if the mean length of the magnetic circuit is l , then

$$Hl = Ni$$

$$H = \frac{Ni}{l}$$

- ⇒ H produces B wherever it exits:

$$B = \mu H$$

$$\mu = \mu_r \mu_o$$

$$\mu_o = \text{permeability of free space} \quad \mu_o = 4\pi \times 10^{-7} \text{ H/m}$$

$$\mu_r = \text{relative permeability}$$

- ⇒ H for air, $\mu_r = 1$; for other magnetic materials $\mu_r = 500-4000$

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- ⇒ for uniform magnetic circuits: X-section area, A ; length l , permeability μ

$$\Lambda = \frac{\phi}{S} = \frac{BA}{HI} = \mu \frac{A}{l}$$

$$S = \frac{1}{\Lambda} = \frac{l}{\mu A}$$

$$\phi = \frac{F}{S} = \frac{Ni}{\mu A}$$

Magnetic circuits	Electric circuits
ϕ	I
F	V
Λ	G
$S = \frac{l}{\mu A}$	$R = \frac{l}{\sigma A}$

- ⇒ in magnetic circuits:

$$F_{\text{source}} = Ni$$

$$F_{\text{used}} = F_{\text{drop}} = \phi S$$

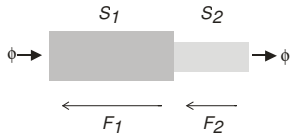
- ⇒ hence:

$$Ni = \phi S$$

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Composite magnetic circuits

I. Series



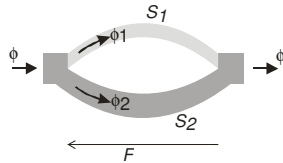
$$F = F_1 + F_2$$

$$\frac{F}{\phi} = \frac{F_1}{\phi} + \frac{F_2}{\phi}$$

$$S_{tot} = S_1 + S_2$$

⇒ reluctances in series and parallel combine together in the same way as resistances

II. Parallel



$$\phi = \phi_1 + \phi_2$$

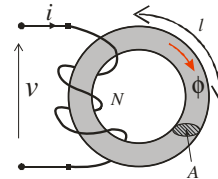
$$\frac{\phi}{F} = \frac{\phi_1}{F} + \frac{\phi_2}{F}$$

$$\frac{1}{S_{tot}} = \frac{1}{S_1} + \frac{1}{S_2}$$

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Inductance

⇒ any circuit in which a change of current causes change of magnetic flux, and hence a change of voltage, is said to have inductance



⇒ definition (elec. circuit point of view):

$$v = L \frac{di}{dt}$$

⇒ Faraday's law:

$$v = N \frac{d\phi}{dt}$$

⇒ equating:

$$L \frac{di}{dt} = N \frac{d\phi}{dt}$$

$$L = N \frac{d\phi}{di}$$

⇒ for uniform case: $L = N \frac{\phi}{i}$

⇒ but $\phi = \frac{F}{S} = \frac{Ni}{S}$

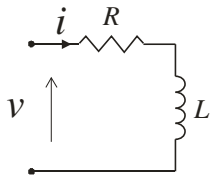
$$L = \frac{N^2}{S}$$

⇒ inductance depends on:

- geometry
- material
- turns

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Inductance on dc



⇒ real coil, with R

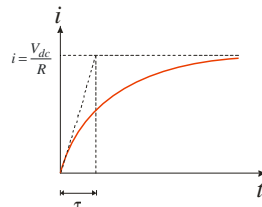
$$v = Ri + L \frac{di}{dt}$$

⇒ on dc

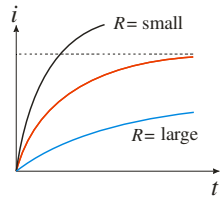
$$V_{dc} = Ri + L \frac{di}{dt}$$

$$i = \frac{V_{dc}}{R} \left(1 - e^{-t/\tau} \right)$$

$$\tau = L/R = \text{time constant}$$



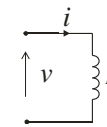
⇒ given L, variable R



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Inductance on ac

⇒ consider pure L



$$v = L \frac{di}{dt}$$

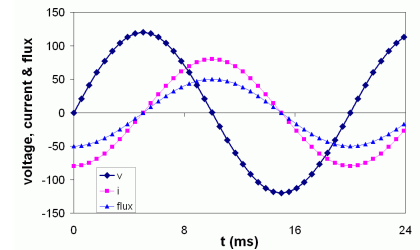
⇒ with sinusoidal supply, v

$$v = V_m \sin \omega t$$

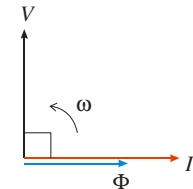
$$i = \frac{1}{L} \int v dt = -\frac{V_m}{\omega L} \cos \omega t + \text{const.}$$

⇒ after a long time, const. = 0

$$\phi = \frac{F}{S} = \frac{Ni}{S} = -\frac{NV_m}{S\omega L} \cos \omega t$$



Phasor diagram



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⇒ for flux:

$$\phi = -\frac{V_m}{\omega N} \cos \omega t$$

$$\Phi_m = \frac{V_m}{\omega N}$$

$$V_m = \omega N \Phi_m$$

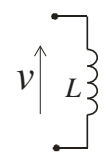
$$\sqrt{2}V = 2\pi f N B_m A$$

$$V = 4.44 f N B_m A \quad \rightarrow \text{the transformer equation}$$

rms value max value

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Real inductor



+ Power Losses:

- Copper loss (R of coil), P_{Cu}
- Iron loss, P_{Fe}
 - Eddy current, P_e
 - Hysteresis, P_h

eddy current

⇒ there is induced voltage in the core of the magnetic material caused by the changing magnetic flux (due to the ac current in coil)

⇒ currents circulating in the core due to induced voltage are called eddy currents

hysteresis

⇒ in the iron core the magnetic flux lags the increase or decrease of the magnetising force; this creates a hysteresis loop

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Iron loss

eddy current loss

⇒ the power loss due to eddy current and the resistance of the iron core

$$P_e = k_e f^2 B_m^2 \quad [W/m^3]$$

hysteresis loss

⇒ the energy loss per cycle obtained by product of the area of the hysteresis loop and the volume of the core material

⇒ 3 factors affect the shape & size of hysteresis loop:

- material: narrow-magnetises easily; wide-does not easily magnetise
- max value of B , $B_m \propto V_m$
- initial magnetising state of iron piece

$$P_h = k_h f B_m^n \quad [W/m^3]$$

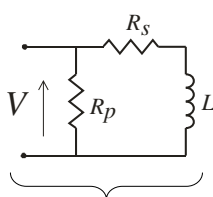
⇒ n = Steinmetz index (1.6-2.5)

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Inductor equivalent circuit

$P_{Cu} = i^2 R \quad \Rightarrow$ represent by resistance, $R_s =$ series resistance

$P_{Fe} \propto \phi^2$
 $\propto V^2 \quad \Rightarrow$ represent by resistance, R_p such that $P_{Fe} = \frac{V^2}{R_p}$
 $= kV^2 \quad \quad \quad k = \frac{1}{R_p}$



⇒ inductance $L + P_{Cu} + P_{Fe}$

inductor

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Equivalent circuit

$$j\omega L = jX_o$$

$$I_{\text{Loss}} = \frac{V_o}{R_o}$$

$$I_{\text{mag}} = \frac{V_o}{jX_o}$$

Phasor diagram

$$\vec{I} = \vec{I}_{\text{Loss}} + \vec{I}_{\text{mag}}$$

$$\vec{V} = \vec{V}_o + \vec{I}R$$

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Limitations of magnetic circuit model

- ⇒ requires homogeneous material
- ⇒ requires isotropic material
- ⇒ neglects non-linear characteristics
- ⇒ neglects saturation
- ⇒ neglects leakage and fringing flux

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Examples:

1) The magnetic circuit shown in the figure has a cross-sectional area everywhere of 100 mm^2 , and permeability of 10^{-3} H/m . If the coil has 100 turns and carries a steady current of 10 A, calculate the magnetic flux in the various portions A, B and C. Assume that the effective length of each side of the squares is 100 mm, and neglect complications at the corners. What is the self-inductance of the coil?

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